Control theory: state-space approach for linear systems Preliminary review

2 janvier 2022

Control theory: state-space approach for linear systems 2 janvier 2022 1/17

(日)

A system (and its model) is said to be

deterministic if for all input u(t), there exists only one possible output y(t). On the contrary, in the non-deterministic or stochastic case, there are several possible outputs, each of them having a probability of occurring.

イロト 不得 トイラト イラト 一日

A system (and its model) is said to be

- deterministic if for all input u(t), there exists only one possible output y(t). On the contrary, in the non-deterministic or stochastic case, there are several possible outputs, each of them having a probability of occurring.
- linear, if the superposition principle applies :

$$\mathcal{S}(k_1 \cdot u_1(t) + k_2 \cdot u_2(t)) = k_1 \cdot \mathcal{S}(u_1(t)) + k_2 \cdot \mathcal{S}(u_2(t)).$$

A system (and its model) is said to be

- deterministic if for all input u(t), there exists only one possible output y(t). On the contrary, in the non-deterministic or stochastic case, there are several possible outputs, each of them having a probability of occurring.
- linear, if the superposition principle applies :

$$\mathcal{S}(k_1 \cdot u_1(t) + k_2 \cdot u_2(t)) = k_1 \cdot \mathcal{S}(u_1(t)) + k_2 \cdot \mathcal{S}(u_2(t)).$$

stationnary if the relations between the input and the output remain the same as time ellapses (no variation of system- parameters).

イロト 不得 トイラト イラト 一日

A system (and its model) is said to be

- deterministic if for all input u(t), there exists only one possible output y(t). On the contrary, in the non-deterministic or stochastic case, there are several possible outputs, each of them having a probability of occurring.
- linear, if the superposition principle applies :

$$\mathcal{S}(k_1 \cdot u_1(t) + k_2 \cdot u_2(t)) = k_1 \cdot \mathcal{S}(u_1(t)) + k_2 \cdot \mathcal{S}(u_2(t)).$$

- stationnary if the relations between the input and the output remain the same as time ellapses (no variation of system- parameters).
- causal if system output at any time instant t₀, y(t₀), does not depend on future values of input u(t), t > t₀ (all the physical systems are causal).

イロト 不得 トイラト イラト 一日



 $\ensuremath{\mathrm{Figure}}$ – container handling gantry crane

We focus on horizontal shifting of set trolley-cables-container (lifting is stopped).

We aim at regulating the container speed while avoiding the container to swing (dangling).

Control theory: state-space approach for linear systems 2 janvier 2022 3 / 17

- Set trolley-cables-container modeled as a pendulum with a mobile pivot
- Cables \equiv indeformable solid, in pivot-connection with the trolley
- Container = punctual mass placed in the center of gravity
- Frictions are neglected



Schématisation de l'ensemble chariot-câbles-container

Abstraction sous la forme d'un pendule

4 日本 4 周本 4 日本 4 日本



т	mass of the container
L	cables length (\equiv indeformable solid)
g	gravity acceleration

p _m ,v _m , a _m	position, speed and acceleration of the mass projected onto axis \overrightarrow{x}
p_c, v_c, a_c	position, speed and acceleration of the trolley projected onto axis \overrightarrow{x}
heta	angle between cables and the vertical axis

イロト イボト イヨト イヨト

э

The goal is to regulate the mass-speed while avoiding it to swing. We consider that :

- the input of the system is the horizontal speed of the trolley v_c (at any time instant t, we assume to be able to set the value of v_c(t));
- the output of the system is the horizontal speed v_m of the mass (container).

We neglect **disturbances (noises)** influencing the system : we could take into account the wind (uncontrolled exogeneous quantity influencing the container-dangling).

This system is **deterministic** : an input v_c leads to only one possible output v_m .

This system is **causal** : the output value at a time instant t_0 , $v_m(t_0)$,does not depend on any future value of the input $v_c(t)$ for $t > t_0$.

Modelling of this system : to obtain the differential equations (parametric model) describing the evolution of variables :

▲□▶ ▲□▶ ▲□▶ ▲□▶ □ ののの

Modelling of this system : to obtain the differential equations (parametric model) describing the evolution of variables :

We have $p_c = p_m + L \sin \theta$, hence the trolley speed is related to the mass-speed by :

$$v_c = v_m + L \frac{d\sin\theta}{dt}.$$
 (1)



Control theory: state-space approach for linear systems 2 janvier 2022 7 / 17

・ロト ・ 同ト ・ ヨト ・ ヨト

r

Fundamental principle of the dynamics (Newton's second law) applied to the mass and projected onto \vec{x} gives : $ma_m = T \sin \theta$. Cables considered as indeformables $\Rightarrow T = mg \cos \theta$. We deduce that

$$ma_m = mg\sin\theta\cos\theta$$
 (2)



System modeled by the equations

$$v_c = v_m + L \frac{d\sin\theta}{dt}$$

$$ma_m = mg\sin\theta\cos\theta$$

- stationnary since the parameters in these differential equations (that is m, g and L) are assumed to be constant as time ellapses.
- not linear.

イロト 不得 トイラト イラト 一日

System modeled by

$$v_c = v_m + L \frac{d\sin\theta}{dt}$$
 $ma_m = mg\sin\theta\cos\theta$

To have a linear model, we assume approximations on the behavior of the system (less precise model but easier to use since linear) : we consider that $\theta \approx 0$ and so $\cos \theta \approx 1$, $\sin \theta \approx \theta$. We deduce the following linear model :

$$\frac{d\theta}{dt} = -\frac{1}{L}v_m + \frac{1}{L}v_c \tag{3}$$

$$\frac{dv_m}{dt}(=a_m) = g\theta \tag{4}$$

イロト 不得 トイラト イラト 二日

Control theory: state-space approach for linear systems 2 janvier 2022 10/17

Reminders about linear systems

	CONTINUOUS TIME	
Convolution	y(t) = (h * u)(t) = $\int_{0}^{t} h(\tau)u(t-\tau)d\tau$	
Transfer F.	$H(s) = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_1 s + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0}$ $\mathcal{L}[h(t)] = H(s)$ Y(s) = H(s)U(s)	
Diff. Equ.	$y^{(n)} + a_{n-1}y^{(n-1)} + \ldots + a_{1}\dot{y} + a_{0}y$ = $b_{m}u^{(m)} + b_{m-1}u^{(m-1)} + \ldots + b_{1}\dot{u} + b_{0}u$ C.I. $y(0^{+}), \dot{y}(0^{+}), \ldots, y^{(n-1)}(0^{+})$	
		-

Control theory: state-space approach for linear systems 2 janvier 2022

11 / 17

Laplace transform of equations

$$\frac{d\theta}{dt} = -\frac{1}{L}v_m + \frac{1}{L}v_c, \qquad \frac{dv_m}{dt}(=a_m) = g\theta,$$

is

$$s\Theta(s) = -rac{1}{L}V_m(s) + rac{1}{L}V_c(s), \qquad sV_m(s) = g\Theta(s),$$

assuming $\dot{\theta}(0) = 0$, $v_m(0) = 0$, $v_c(0) = 0$ for initial conditions.

A transfer function can then be deduced.

We have obtained :

$$\begin{cases} s\Theta(s) = -\frac{1}{L}V_m(s) + \frac{1}{L}V_c(s) \\ sV_m(s) = g\Theta(s) \Longleftrightarrow \Theta(s) = \frac{s}{g}V_m(s) \end{cases}$$

This leads to

$$rac{s^2}{g}V_m(s) = -rac{1}{L}V_m(s) + rac{1}{L}V_c(s)$$
 $\iff \left(rac{s^2}{g} + rac{1}{L}
ight)V_m(s) = rac{1}{L}V_c(s)$

Hence

$$\frac{V_m(s)}{V_c(s)} = \frac{1}{1+\frac{L}{g}s^2} = \frac{Y(s)}{U(s)} = H(s).$$

Control theory: state-space approach for linear systems 2 janvier 2022 13 / 17

イロト イヨト イヨト イヨト 二日

The obtained transfer function

$$H(s) = rac{Y(s)}{U(s)} = rac{V_m(s)}{V_c(s)} = rac{1}{1 + rac{L}{g}s^2}$$

is a second-order ones whose standard form is

$$H(s) = K rac{1}{1+2\xi rac{s}{\omega_n}+(rac{s}{\omega_n})^2}$$

By identification, we can deduce $\omega_n = \sqrt{\frac{g}{L}}$ and $\xi = 0$.

Control theory: state-space approach for linear systems 2 janvier 2022 14 / 17

▲□▶ ▲□▶ ▲ □▶ ▲ □▶ ▲ □ ● のへで



A linear system is said to be *stable* if for all bounded-range input, the output-response has also a bounded range.

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ - 三 - のへで



A linear system is said to be *stable* if for all bounded-range input, the output-response has also a bounded range.

Equivalently, a linear system is said to be *stable* if when the input is set to zero, then the output-response tends to zero while t tends to infinity.

イロト 不得 トイヨト イヨト 二日



A linear system is said to be *stable* if for all bounded-range input, the output-response has also a bounded range.

Equivalently, a linear system is said to be *stable* if when the input is set to zero, then the output-response tends to zero while t tends to infinity.

Theorem (Stability criterion regarding the transfer function)

Let a system be described by its transfer function H(s). It is stable if, and only if, the poles (i.e. the roots of the denominator of H(s)) have all a strictly negative real part.

イロト 不得下 イヨト イヨト 二日

We can guess that the system corresponding to the set trolley-cables-container is not stable. For example, if the input is first set to a non-zero given value (the trolley-speed is set to a non-zero value), and then set to zero (the trolley is stopped) : in that case we can expect that the container will swing and these oscillations will continue indefinitely (frictions are neglected so that the range of the oscillations will not decrease) which means that the output-response will not tend to zero asymptotically.

We can guess that the system corresponding to the set trolley-cables-container is not stable. For example, if the input is first set to a non-zero given value (the trolley-speed is set to a non-zero value), and then set to zero (the trolley is stopped) : in that case we can expect that the container will swing and these oscillations will continue indefinitely (frictions are neglected so that the range of the oscillations will not decrease) which means that the output-response will not tend to zero asymptotically.

$$\mathcal{H}(s) = rac{1}{1+rac{L}{g}s^2}$$

The roots of $1 + \frac{L}{g}s^2$ (denominator of H(s), the transfer function) are $\pm i\sqrt{\frac{g}{L}}$. Their real-part is not strictly negative. This demonstrates that the system is unstable.

A ロ ト 4 回 ト 4 三 ト 4 三 ト 9 0 0 0

Precision

A controlled-system with output y(t) is all the more precise that the difference between the desired output $y_d(t)$ and the actual output y(t) is low. The *precision* can be quantified by :

$$\varepsilon(t) \triangleq y_d(t) - y(t)$$

Precision

A controlled-system with output y(t) is all the more precise that the difference between the desired output $y_d(t)$ and the actual output y(t) is low. The *precision* can be quantified by :

$$\varepsilon(t) riangleq y_d(t) - y(t)$$

The static precision ε_0 denoting this error $\varepsilon(t)$ when t tends to infinity (steady-state error), that is :

$$\varepsilon_0 = \lim_{t \to \infty} \varepsilon(t)$$
.

▲□▶ ▲□▶ ▲□▶ ▲□▶ □ ののの

Precision

A controlled-system with output y(t) is all the more precise that the difference between the desired output $y_d(t)$ and the actual output y(t) is low. The *precision* can be quantified by :

$$\varepsilon(t) riangleq y_d(t) - y(t)$$

The static precision ε_0 denoting this error $\varepsilon(t)$ when t tends to infinity (steady-state error), that is :

$$\varepsilon_0 = \lim_{t\to\infty} \varepsilon(t)$$
.

Sometimes, the *stationary error* is considered. The stationary error of order n, denoted ε_n , is the steady-state error for an input of the form $U(s) = \frac{1}{s^n}$.